appropriately modified.⁷ Furthermore, since there is no phase shift at a turning point unless the turning point lies on a caustic ([remarks after eq. (12)],⁴ the phase factors $\exp(\pm ks\pi/2)$ in Y_{path} [cf. (15.3), (15.9)],¹ and subsequently where Y_{path} appears in our results) are replaced by unity except at caustics. In other respects our earlier results remain unmodified.

ACKNOWLEDGMENT

The author wishes to thank Prof. Claude W. Horton, Sr., University of Texas and Dr. Thomas G. Kincaid, General Electric Research Division, for their helpful comments.

⁷ D. Middleton, "Probability models of received scattered and ambient fields," in *Ocean 72*, *IEEE Int. Symp. Engineering in Ocean Environment*, pp. 8-14, 1972.

Sharper Lower Bounds for Discrimination Information in Terms of Variation

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Abstract—The lower bound for discrimination information in terms of variation, derived recently by Kullback [7] for the distribution-free case, is sharpened. Furthermore, under a restriction, a lower bound is derived that is sharper than all other existing bounds.

Given two probability distributions $f_1(x)$ and $f_2(x)$, there are two well-known measures of the "distance" or difference between $f_1(x)$ and $f_2(x)$. One is the discrimination information given by

$$I = \int f_1(x) \log \left[\frac{f_1(x)}{f_2(x)} \right] dx. \tag{1}$$

The other is the variation given by

$$V = \int |f_1(x) - f_2(x)| dx.$$
 (2)

In the past there has been a great deal of interest in bounding I in terms of V. Volkonskij and Rozanov [1] showed that

$$I \ge V - \log(1 + V). \tag{3}$$

Pinsker [2] improved (3) by showing that

$$I \ge \frac{V^2}{\Gamma},$$
 (4)

where Γ is a constant greater than two. Csiszár [3] proved (4) with $\Gamma = 16$ while McKean [4] established (4) with $\Gamma = 4e$. Csiszár [5] and Kemperman [6], apparently independently, sharpened these results by proving that

$$I \ge \frac{V^2}{2}. (5)$$

Kullback [7], [8] sharpened (5) by showing that

$$I \ge \frac{V^2}{2} + \frac{V^4}{36}. (6)$$

The disadvantage of the bounds (3)-(6) is that for V close to two they are loose, and for V=2 the equality does not hold. In an attempt to improve these bounds, at least for V close to

two, Vajda [9] proved that

$$I \ge \log\left(\frac{2+V}{2-V}\right) - \frac{2V}{2+V}.\tag{7}$$

The bound given by (7) is slightly looser than (6) for V less than approximately 1.75, but much sharper than (6) for $V \gtrsim 1.75$. Furthermore, it has the added nice property that the equality holds for both V = 0 and V = 2.

In this correspondence Kullback's bound (6) is sharpened further. In fact, it will be shown that

$$I \ge \frac{V^2}{2} + \frac{V^4}{36} + \frac{V^6}{288}. (8)$$

Thus the maximum of (8) and (7) provides the sharpest lower bound available for I in terms of V for arbitrary distributions.

Let Ω_i denote the space where $f_i(x) > f_j(x)$, $i = 1, 2, i \neq j$. For the set of distribution pairs such that

$$\int_{\Omega_1} f_2(x) \ dx = \int_{\Omega_2} f_1(x) \ dx, \tag{9}$$

which holds, for example, for the important case of Gaussian distributions with equal covariance matrices, it will be shown that

$$I \ge \frac{V}{2} \log \left(\frac{2+V}{2-V} \right),\tag{10}$$

where the equality holds for both V = 0 and V = 2. Furthermore, it will be shown that (10) is sharper than both (7) and (8), for every $V \in [0,2]$.

Proof of (8): Let L(u,t) be a function given by

$$L(u,t) = (u + t) \log \left(1 + \frac{t}{u}\right) + (1 - u - t) \log \left(1 - \frac{t}{1 - u}\right), \quad (11)$$

where u and t are real numbers. Also let

$$\int_{\Omega_1} f_2(x) \ dx = \alpha_2 \tag{12}$$

and

$$\int_{\Omega_2} f_1(x) \ dx = \alpha_1. \tag{13}$$

It follows from (12) and (13), see [11], that

$$V = 2(1 - \alpha_1 - \alpha_2). \tag{14}$$

Krafft and Schmitz [10] showed that for

$$0 < u < 1 \tag{15}$$

and

$$-u < t < 1 - u \tag{16}$$

it holds that

$$L(u,t) \ge 2t^2 + \frac{4t^4}{9} + \frac{2t^6}{9}. \tag{17}$$

It is easy to verify that for $u = 1 - \alpha_2$ and $t = \alpha_1 + \alpha_2 - 1$, (15) and (16) are satisfied and

$$L(1 - \alpha_2, \alpha_1 + \alpha_2 - 1) = \alpha_1 \log \left(\frac{\alpha_1}{1 - \alpha_2}\right) + (1 - \alpha_1) \log \left(\frac{1 - \alpha_1}{\alpha_2}\right). \quad (18)$$

Manuscript received April 4, 1974; revised June 26, 1974. The author is with the School of Computer Science, McGill University, Montreal, P.Q., Canada.

Kullback [12] has shown that

$$I \ge L(1 - \alpha_2, \alpha_1 + \alpha_2 - 1).$$
 (19)

It also follows from (14) that

$$t = -\frac{V}{2}. (20)$$

Substituting (20) into (17) and combining the latter with (19) yields (8), the desired result.

Proof of (10): Substitute $\alpha_1 = \alpha_2 = \frac{1}{2} - V/4$ into (19). Equation (10) can be used to obtain useful bounds in pattern recognition [13]. To prove that (10) is sharper than all the other bounds in this correspondence, for all $V \in [0,2]$, it is sufficient to prove that (10) is sharper than (7) and (8).

Substitute $u = \frac{1}{2} + V/4$ and t = -V/2 into (17). Then the left side of (17) becomes the right side of (10), and the right side of (17) becomes the right side of (8), thus proving that (10) is

In order to prove that (10) is sharper than (7) it must be shown that

$$\frac{2x}{1-x^2}-\log\left(\frac{1+x}{1-x}\right)\geq 0,$$

where x = V/2, which follows from the fact that, for $0 \le x \le 1$, we have

$$\frac{1}{2}\log\left(\frac{1+x}{1-x}\right) = \int_0^x (1-y^2)^{-1} dy$$

$$\leq (1-x^2)^{-1} \int_0^x dy = \frac{x}{1-x^2}.$$

ACKNOWLEDGMENT

The author is indebted to one of the referees for suggesting some of the preceding proofs, which are simpler than the original

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The Capacity Region of a Multiple-Access Discrete Memoryless Channel Can Increase with Feedback

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Abstract—The capacity of a single-input single-output discrete memoryless channel is not increased by the use of a noiseless feedback link. It is shown, by example, that this is not the case for a multipleaccess discrete memoryless channel. That is, it is shown that the capacity region for such a channel is enlarged if a noiseless feedback link is utilized.

Introduction

Shannon [1] proved that the capacity of a single-input singleoutput discrete memoryless channel is not increased even if the encoder could observe the output of the channel via a noiseless delayless feedback link. Recently, Liao [2], and then, Slepian and Wolf [3] gave formulas for the capacity region of a two-input single-output discrete memoryless channel with independent encoding of two source messages. After summarizing their results, we evaluate the performance of a transmission scheme for this channel, which makes use of noiseless feedback links from the output to the two encoders. We show that this scheme yields a vanishingly small error probability for a pair of rates that lies outside the capacity region.

CAPACITY REGIONS WITHOUT FEEDBACK

In this section we summarize the previously published results concerned with the capacity region of a multiple-access discrete memoryless channel without feedback. Consider the block diagram shown in Fig. 1. Two sources are described by a twodimensional rate vector $\mathbf{R} = (R_1, R_2)$ with nonnegative components. Let N be a fixed positive integer. Every N time units, the sources¹ produce a pair of statistically independent random variables (U_1, U_2) , where U_i is uniformly distributed over the set of integers $\{1, 2, \dots, M_i = \lceil 2^{R_i N} \rceil\} \triangleq \mathcal{I}_i, i = 1,2$. Here [x] is the smallest integer greater than or equal to x.

The channel is described by a conditional probability distribution of the output random variable Y (which takes values $y \in \mathcal{Y}$) given the inputs $X_1 = x_1 \in \mathcal{X}_1$ and $X_2 = x_2 \in \mathcal{X}_2$. We denote this conditional probability distribution $P_{Y|X_1X_2}(y \mid x_1x_2)$. The channel is assumed memoryless in the usual sense. That is, the conditional probability distribution for N-vectors is equal to the product of the marginal conditional probability distributions. The encoders are a pair of deterministic mappings from the source outputs to channel input N-vectors. The mappings are such that if the sources produce the pair $(U_1 = i, U_2 = j)$, encoder 1 produces the N-vector $x_{1i} \in (\mathcal{X}_1)^N$, which depends only on i, and encoder 2 produces the N-vector $x_{2j} \in (\mathcal{X}_2)^N$, which depends only on j.

The decoder is a deterministic mapping from the channel output N-vector y to the pair (i^*, j^*) , where $i^* \in \mathcal{I}_1$, $j^* \in \mathcal{I}_2$. We denote the decoder outputs by the pair of random variables $(U_1^*, U_2^*).$

Manuscript received April 10, 1973; revised July 5, 1974. This work was supported in part by the U.S. Air Force Office of Scientific Research under Contract F44620-72C-0085.

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¹ Henceforth, source does not refer to the actual source with rate R_i ; it refers to an extended source with the larger rate N^{-1} log $[2R_iN]$, which is compatible with the block length N. This extended source consists of the actual source and additional devices to add bits when necessary.